Phase Locked Loop

The MC14046B phase locked loop contains two phase comparators, a voltage-controlled oscillator (VCO), source follower, and zener diode. The comparators have two common signal inputs, PCAin and PCBin. Input PCAin can be used directly coupled to large voltage signals, or indirectly coupled (with a series capacitor) to small voltage signals. The self-bias circuit adjusts small voltage signals in the linear region of the amplifier. Phase comparator 1 (an exclusive OR gate) provides a digital error signal PC1_{OUT}, and maintains 90° phase shift at the center frequency between PCAin and PCBin signals (both at 50% duty cycle). Phase comparator 2 (with leading edge sensing logic) provides digital error signals, PC2out and LD, and maintains a 0° phase shift between PCAin and PCBin signals (duty cycle is immaterial). The linear VCO produces an output signal VCO_{OUT} whose frequency is determined by the voltage of input VCOin and the capacitor and resistors connected to pins C1_A, C1_B, R1, and R2. The source-follower output SFout with an external resistor is used where the VCOin signal is needed but no loading can be tolerated. The inhibit input Inh, when high, disables the VCO and source follower to minimize standby power consumption. The zener diode can be used to assist in power supply regulation.

Applications include FM and FSK modulation and demodulation, frequency synthesis and multiplication, frequency discrimination, tone decoding, data synchronization and conditioning, voltage-to-frequency conversion and motor speed control.

- Buffered Outputs Compatible with MHTL and Low–Power TTL
- Diode Protection on All Inputs
- Supply Voltage Range = 3.0 to 18 V
- Pin-for-Pin Replacement for CD4046B
- Phase Comparator 1 is an Exclusive Or Gate and is Duty Cycle Limited
- Phase Comparator 2 switches on Rising Edges and is not Duty Cycle
 Limited



L SUFFIX
CERAMIC
CASE 620P SUFFIX
CASE 648DW SUFFIX
SOIC

MC14046B

ORDERING INFORMATION

MC14XXXBCP MC14XXXBCL MC14XXXBDW

Plastic Ceramic SOIC

CASE 751G

 $T_A = -55^\circ$ to 125°C for all packages.

PIN ASSIGNMENT							
LD [1•	16					
PC1 _{out}	2	15] ZENER				
PCB _{in} [3	14] PCA _{in}				
VCO _{out} [4	13] PC2 _{out}				
ілн [5	12] R2				
С1 _А [6	11] R1				
С1 _В [7	10] SF _{out}				
v _{ss} [8	9] ∨CO _{in}				
-							



10/97

MAXIMUM RATINGS* (Voltages Referenced to VSS)

Rating	Symbol	Value	Unit
DC Supply Voltage	V _{DD}	– 0.5 to + 18	Vdc
Input Voltage, All Inputs	V _{in}	– 0.5 to V _{DD} + 0.5	Vdc
DC Input Current, per Pin	l _{in}	± 10	mAdc
Power Dissipation, per Package†	PD	500	mW
Operating Temperature Range	т _А	– 55 to + 125	°C
Storage Temperature Range	T _{stg}	– 65 to + 150	°C

* Maximum Ratings are those values beyond which damage to the device may occur.

†Temperature Derating:

Plastic "P and D/DW" Packages: - 7.0 mW/°C From 65°C To 125°C Ceramic "L" Packages: - 12 mW/°C From 100°C To 125°C

ELECTRICAL CHARACTERISTICS (Voltages Referenced to VSS)

			VDD	- 5	5°C		25°C		125	5°C	
Characteristic		Symbol	Vdc	Min	Max	Min	Тур	Max	Min	Max	Unit
Output Voltage V _{in} = V _{DD} or 0	"0" Level	VOL	5.0 10 15		0.05 0.05 0.05		0 0 0	0.05 0.05 0.05		0.05 0.05 0.05	Vdc
V _{in} = 0 or V _{DD}	"1" Level	VOH	5.0 10 15	4.95 9.95 14.95		4.95 9.95 14.95	5.0 10 15		4.95 9.95 14.95		Vdc
Input Voltage # $(V_O = 4.5 \text{ or } 0.5 \text{ Vdc})$ $(V_O = 9.0 \text{ or } 1.0 \text{ Vdc})$ $(V_O = 13.5 \text{ or } 1.5 \text{ Vdc})$	"0" Level	VIL	5.0 10 15		1.5 3.0 4.0		2.25 4.50 6.75	1.5 3.0 4.0		1.5 3.0 4.0	Vdc
$(V_O = 0.5 \text{ or } 4.5 \text{ Vdc})$ $(V_O = 1.0 \text{ or } 9.0 \text{ Vdc})$ $(V_O = 1.5 \text{ or } 13.5 \text{ Vdc})$	"1" Level	VIH	5.0 10 15	3.5 7.0 11	 	3.5 7.0 11	2.75 5.50 8.25		3.5 7.0 11		Vdc
$\begin{array}{l} \mbox{Output Drive Current} \\ (V_{OH} = 2.5 \mbox{ Vdc}) \\ (V_{OH} = 4.6 \mbox{ Vdc}) \\ (V_{OH} = 9.5 \mbox{ Vdc}) \\ (V_{OH} = 13.5 \mbox{ Vdc}) \end{array}$	Source	IOH	5.0 5.0 10 15	- 1.2 - 0.25 - 0.62 - 1.8	 	- 1.0 - 0.2 - 0.5 - 1.5	- 1.7 - 0.36 - 0.9 - 3.5	 	- 0.7 - 0.14 - 0.35 - 1.1	 	mAdc
(V _{OL} = 0.4 Vdc) (V _{OL} = 0.5 Vdc) (V _{OL} = 1.5 Vdc)	Sink	IOL	5.0 10 15	0.64 1.6 4.2		0.51 1.3 3.4	0.88 2.25 8.8		0.36 0.9 2.4		mAdc
Input Current		l _{in}	15	—	± 0.1	—	±0.00001	± 0.1	—	± 1.0	μAdc
Input Capacitance		C _{in}	—	—	—	—	5.0	7.5	—	—	pF
Quiescent Current (Per Package) Inh = PC/ Zener = VCO _{in} = 0 V, PC or 0 V, I _{out} = 0 µA		I _{DD}	5.0 10 15		5.0 10 20	 	0.005 0.010 0.015	5.0 10 20	 	150 300 600	μAdc
Total Supply Current† (Inh = "0", $f_0 = 10$ kHz, C R1 = 1.0 M Ω , R2 = ∞ Rg and 50% Duty Cycle)	_	ŀΤ	5.0 10 15			I _T = (2	.46 μA/kHz) .91 μA/kHz) .37 μA/kHz)	f + I _{DD}	•		mAdc

#Noise immunity specified for worst-case input combination.

Noise Margin for both "1" and "0" level = 1.0 Vdc min @ V_{DD} = 5.0 Vdc 2.0 Vdc min @ $V_{DD} = 10$ Vdc 2.5 Vdc min @ $V_{DD} = 15$ Vdc

†To Calculate Total Current in General:

$$\begin{split} I_T &\approx 2.2 \ x \ V_{DD} \Big(\frac{VCO_{in} - 1.65}{R1} + \frac{V_{DD} - 1.35}{R2} \Big)^{3/4} + 1.6 \ x \Big(\frac{VCO_{in} - 1.65}{R_{SF}} \Big)^{3/4} + 1 \ x \ 10^{-3} \ (C_L + 9) \ V_{DD} \ f + 1 \ x \ 10^{-1} \ V_{DD}^2 \ \Big(\frac{100\% \ \text{Duty Cycle of PCA}_{in}}{100} \Big) + I_Q \qquad \text{where:} \ I_T \ \text{in } \mu\text{A}, \ C_L \ \text{in } p\text{F}, \ VCO_{in}, \ V_{DD} \ \text{in } Vdc, \ f \ in \ k\text{Hz}, \ \text{and} \ R1, \ R2, \ R_{SF} \ \text{in } M\Omega, \ C_L \ on \ VCO_{out}. \end{split}$$

ELECTRICAL CHARACTERISTICS* ($C_L = 50 \text{ pF}, T_A = 25^{\circ}C$)

		VDD	Minimum		Maximum	
Characteristic	Symbol	Vdc	Device	Typical	Device	Units
Output Rise Time	ttlh					ns
t _{TLH} = (3.0 ns/pF) C _L + 30 ns		5.0	-	180	350	
t _{TLH} = (1.5 ns/pF) C _L + 15 ns		10	-	90	150	
t _{TLH} = (1.1 ns/pF) C _L + 10 ns		15	—	65	110	
Output Fall Time	^t THL					ns
t _{THL} = (1.5 ns/pF) C _L + 25 ns		5.0	-	100	175	
t _{THL} = (0.75 ns/pF) C _L + 12.5 ns		10	-	50	75	
t _{THL} = (0.55 ns/pF) C _L + 9.5 ns		15	_	37	55	
PHASE COMPARATORS 1 and 2						
Input Resistance — PCA _{in}	R _{in}	5.0	1.0	2.0	—	MΩ
		10	0.2	0.4	-	
		15	0.1	0.2	—	
— PCB _{in}	R _{in}	15	150	1500	—	MΩ
Minimum Input Sensitivity	V _{in}	5.0	-	200	300	mV p–p
AC Coupled — PCA _{in}		10	-	400	600	
C series = 1000 pF, f = 50 kHz		15	-	700	1050	
DC Coupled — PCA _{in} , PCB _{in}	—	5 to 15	See	e Noise Immu	nity	
VOLTAGE CONTROLLED OSCILLATOR (VCO)						
Maximum Frequency	fmax	5.0	0.5	0.7	—	MHz
(VCO _{in} = V _{DD} , C1 = 50 pF		10	1.0	1.4	l —	
$R1 = 5.0 \text{ k}\Omega$, and $R2 = \infty$)		15	1.4	1.9	—	
Temperature — Frequency Stability	_	5.0	_	0.12	_	%/°C
(R2 = ∞)		10	_	0.04	_	
		15	-	0.015	_	
Linearity (R2 = ∞)	_					%
(VCO _{in} = 2.5 V ± 0.3 V, R1 > 10 kΩ)		5.0	_	1.0	—	
(VCO _{in} = 5.0 V ± 2.5 V, R1 > 400 kΩ)		10	l —	1.0	—	
$(\text{VCO}_{\text{in}} = 7.5 \text{ V} \pm 5.0 \text{ V}, \text{ R1} \ge 1000 \text{ k}\Omega)$		15	—	1.0	—	
Output Duty Cycle	-	5 to 15	-	50	—	%
Input Resistance — VCO _{in}	R _{in}	15	150	1500	—	MΩ
SOURCE-FOLLOWER	-					
Offset Voltage		5.0	_	1.65	2.2	V
(VCO _{in} minus SF _{out} , RSF > 500 kΩ)		10	-	1.65	2.2	
		15	-	1.65	2.2	
Linearity						%
$(VCO_{in} = 2.5 \text{ V} \pm 0.3 \text{ V}, \text{ R}_{SF} > 50 \text{ k}\Omega)$		5.0	-	0.1	-	
$(VCO_{in} = 5.0 \text{ V} \pm 2.5 \text{ V}, \text{ R}_{SF} > 50 \text{ k}\Omega)$		10	-	0.6	-	
(VCO _{in} = 7.5 V \pm 5.0 V, R _{SF} > 50 k Ω)		15	_	0.8	-	
ZENER DIODE						
Zener Voltage (I _Z = 50 μ A)	VZ	—	6.7	7.0	7.3	V
Dynamic Resistance (I _Z = 1.0 mA)	RZ	_	-	100	_	Ω
The formula given is for the typical characteristics only.			•		-	•

* The formula given is for the typical characteristics only.



Refer to Waveforms in Figure 3.

Figure 1. Phase Comparators State Diagrams

Characteristic	Using Phase Comparator 1	Using Phase Comparator 2			
No signal on input PCA _{in} .	VCO in PLL system adjusts to center frequency (f_0).	VCO in PLL system adjusts to minimum frequency (f _{min}).			
Phase angle between PCA _{in} and PCB _{in} .	90° at center frequency (f_0), approaching 0° and 180° at ends of lock range (2f_)	Always 0° in lock (positive rising edges).			
Locks on harmonics of center frequency.	Yes	No			
Signal input noise rejection.	High	Low			
Lock frequency range (2fL).	The frequency range of the input signal on which the loop will stay locked if it was initially in lock; $2f_L = full VCO$ frequency range = $f_{max} - f_{min}$.				
Capture frequency range (2f _C).	The frequency range of the input signal on which the loop will lock if it was initially out of lock.				
	Depends on low–pass filter characteristics (see Figure 3). f_C \leq f_L	$f_{C} = f_{L}$			
Center frequency (f ₀).	The frequency of VCO _{out} , when VCO _{in} = $1/2$	VDD			
VCO output frequency (f).	$f_{min} = \frac{1}{R_2(C_1 + 32 \text{ pF})} \qquad (V_{CO} \text{ input} = V_{SS})$				
Note: These equations are intended to be a design guide. Since calculated component values may be in error by as much as a factor of 4, laboratory experimentation may be required for fixed designs. Part to part frequency variation with identical passive components is typically less than \pm 20%.	$f_{max} = \frac{1}{R_1(C_1 + 32 \text{ pF})} + f_{min} (V_{CO})$ Where: 10K $\leq R_1 \leq 1 \text{ M}$ 10K $\leq R_2 \leq 1 \text{ M}$ 100pF $\leq C_1 \leq .01 \mu\text{F}$	D inbnt = ADD)			



NOTE: Sometimes R3 is split into two series resistors each R3 ÷ 2. A capacitor C_C is then placed from the midpoint to ground. The value for C_C should be such that the corner frequency of this network does not significantly affect ω_n . In Figure B, the ratio of R3 to R4 sets the damping, R4 \cong (0.1)(R3) for optimum results.

D (1)/	N	Filter A	Filter B
Definitions:	N = Total division ratio in feedback loop $K\phi = V_{DD}/\pi$ for Phase Comparator 1 $K\phi = V_{DD}/4 \pi$ for Phase Comparator 2 $K_{VCO} = \frac{2 \pi \Delta f_{VCO}}{V_{DD} - 2 V}$	$\omega_{\rm n} = \sqrt{\frac{{\rm K}_{\phi}{\rm K}{\rm VCO}}{{\rm NR}_{3}{\rm C}_{2}}}$	$\omega_{n} = \sqrt{\frac{K\varphi KVCO}{NC_{2}(R_{3} + R_{4})}}$
	for a typical design $\omega_n \approx \frac{2 \pi f_r}{10}$ (at phase detector input)	$\zeta = \frac{N\omega_n}{2K_\varphi K_{VCO}}$	$\zeta = 0.5 \omega_{\text{n}} (\text{R}_{3}\text{C}_{2} + \frac{\text{N}}{\text{K}_{\phi}\text{K}_{\text{VCO}}})$
	$\zeta \cong 0.707$	$F(s) = \frac{1}{R_3C_2S + 1}$	$F(s) = \frac{R_3C_2S + 1}{S(R_3C_2 + R_4C_2) + 1}$







LOW-PASS FILTER

Note: for further information, see:

(1) F. Gardner, "Phase–Lock Techniques", John Wiley and Son, New York, 1966.

(2) G. S. Moschytz, "Miniature RC Filters Using Phase-Locked Loop", BSTJ, May, 1965.

- (3) Garth Nash, "Phase–Lock Loop Design Fundamentals", AN–535, Motorola Inc.
- (4) A. B. Przedpelski, "Phase-Locked Loop Design Articles", AR254, reprinted by Motorola Inc.

Figure 3. General Phase–Locked Loop Connections and Waveforms

OUTLINE DIMENSIONS



OUTLINE DIMENSIONS



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